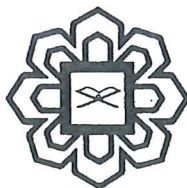


Name: \_\_\_\_\_

Matric No: \_\_\_\_\_ Section: \_\_\_\_\_



الجامعة الإسلامية العالمية ماليزيا

INTERNATIONAL ISLAMIC UNIVERSITY MALAYSIA

MID-TERM EXAMINATION  
SEMESTER II, 2017/2018 SESSION  
KULLIYAH OF ENGINEERING

Programme : ENGINEERING Level of Study : UG 2  
Time : 5:15 pm-7:15 pm Date : 05/04/2018  
Duration : 2 Hours  
Course Code : EECE 2313/ ECE 2133 Section(s) : 1, 2  
Course Title : **Electronic Circuits**

This Question Paper consists of Five (5) Printed Pages (Including Cover and a blank page) with Three (3) Questions.

**INSTRUCTION(S) TO CANDIDATES**

DO NOT OPEN UNTIL YOU ARE ASKED TO DO SO

- Use only pen for writing answer.
- Do not use your own sheet.
- A total mark of this examination is 60.
- This examination is worth 30% of the total assessment.
- For drawing you may use pencil
- Answer **ALL THREE(3)** questions.
- Answer on the question paper.

**Any form of cheating or attempt to cheat is a serious offence which may lead to dismissal.**

	Question 1	Question 2	Question 3	Total Marks
Marks	20	20	20	60
Marks Obtained				

## Q.1 [20 marks]

- (a) Consider the circuit as shown in Fig. 1(a), derive the expression (step by step) for the voltage transfer function  $T(s) = v_o(s)/v_i(s)$ . Find the corner frequency of the circuit if  $R_s = 3 \text{ k}\Omega$ ,  $R_p = 9 \text{ k}\Omega$  and  $C_{p1} = 8 \text{ nF}$  and  $C_{p2} = 8 \text{ nF}$ . (6+2 marks)

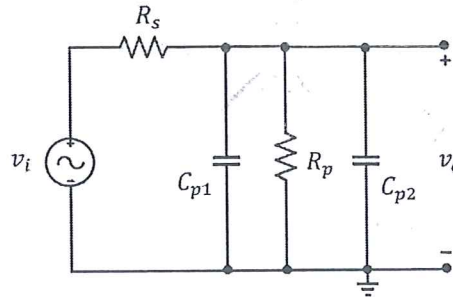
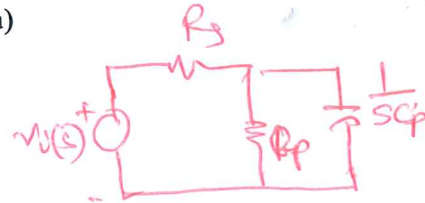


Fig. 1(a)

$$C_p = C_{p1} + C_{p2}$$



$$Z_p = R_p \parallel \left(\frac{1}{sC_p}\right)$$

$$= \frac{R_p \times \frac{1}{sC_p}}{R_p + \frac{1}{sC_p}} = \frac{R_p}{1 + sR_pC_p}$$

$$Z_p + Z_s = \frac{R_p}{1 + sR_pC_p} + R_s = \frac{(R_p + R_s) + sR_pR_sC_p}{1 + sR_pC_p}$$

$$\therefore T(s) = \frac{V_o(s)}{V_i(s)} = \frac{Z_p}{Z_p + Z_s} = \frac{\frac{R_p}{1 + sR_pC_p}}{\frac{(R_p + R_s) + sR_pR_sC_p}{1 + sR_pC_p}} = \frac{R_p}{R_p + R_s + sC_pR_pR_s}$$

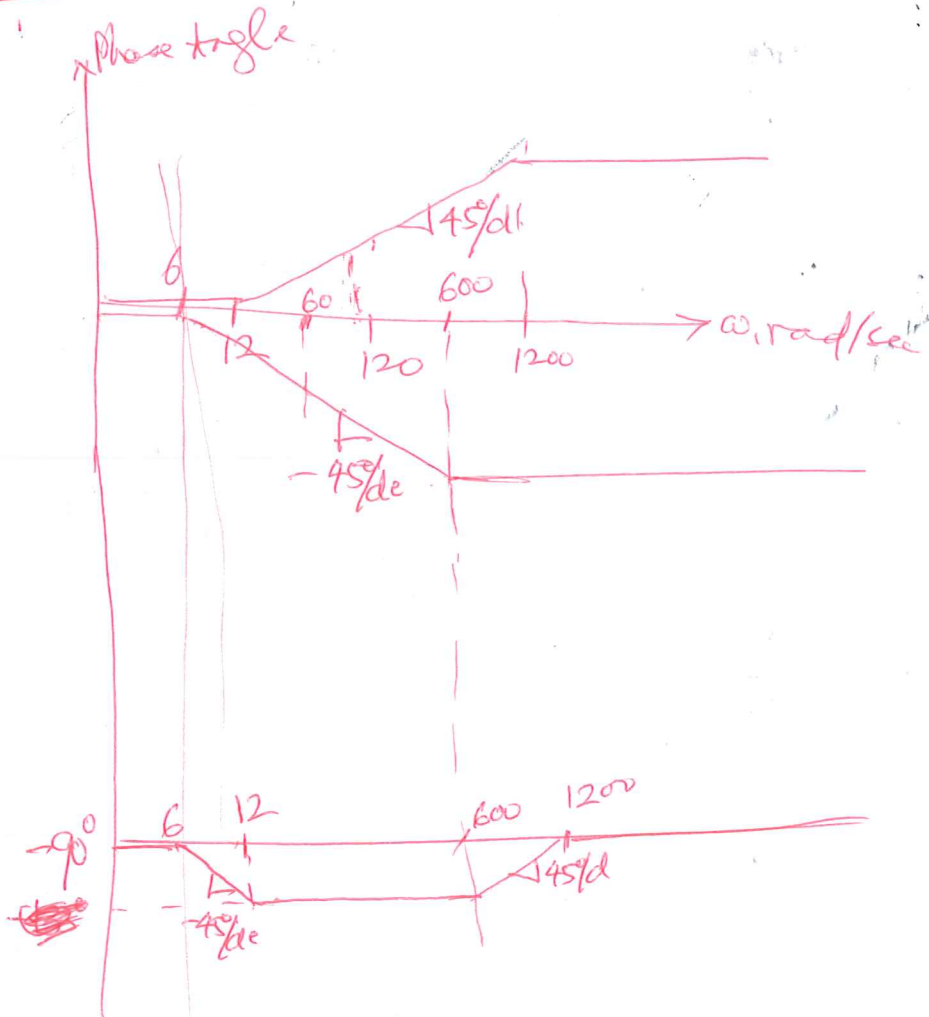
$$= \frac{R_p}{R_p + R_s} \cdot \frac{1}{1 + s \frac{R_pR_s}{R_p + R_s} C_p} = \frac{R_p}{R_p + R_s} \cdot \frac{1}{1 + s\tau_c}$$

$$\tau_c = R_p \parallel R_s \times C_p = \left(\frac{3 \parallel 9}{k}\right) \times (8 + 8) \text{ nF}$$

$$= 2.5 \text{ k} \times 16 \text{ nF} = 3.6 \times 10^{-5} \text{ sec} = 36 \mu\text{s}$$

$$f_c = \frac{1}{2\pi\tau_c} = 4.42 \text{ kHz} \quad \leftarrow$$

Phase.



$$\theta = -90^\circ + \tan^{-1} \frac{\omega}{120} - \tan^{-1} \frac{\omega}{60}$$
$$= -90^\circ + 39.8^\circ - 59.036^\circ = -109.23^\circ$$

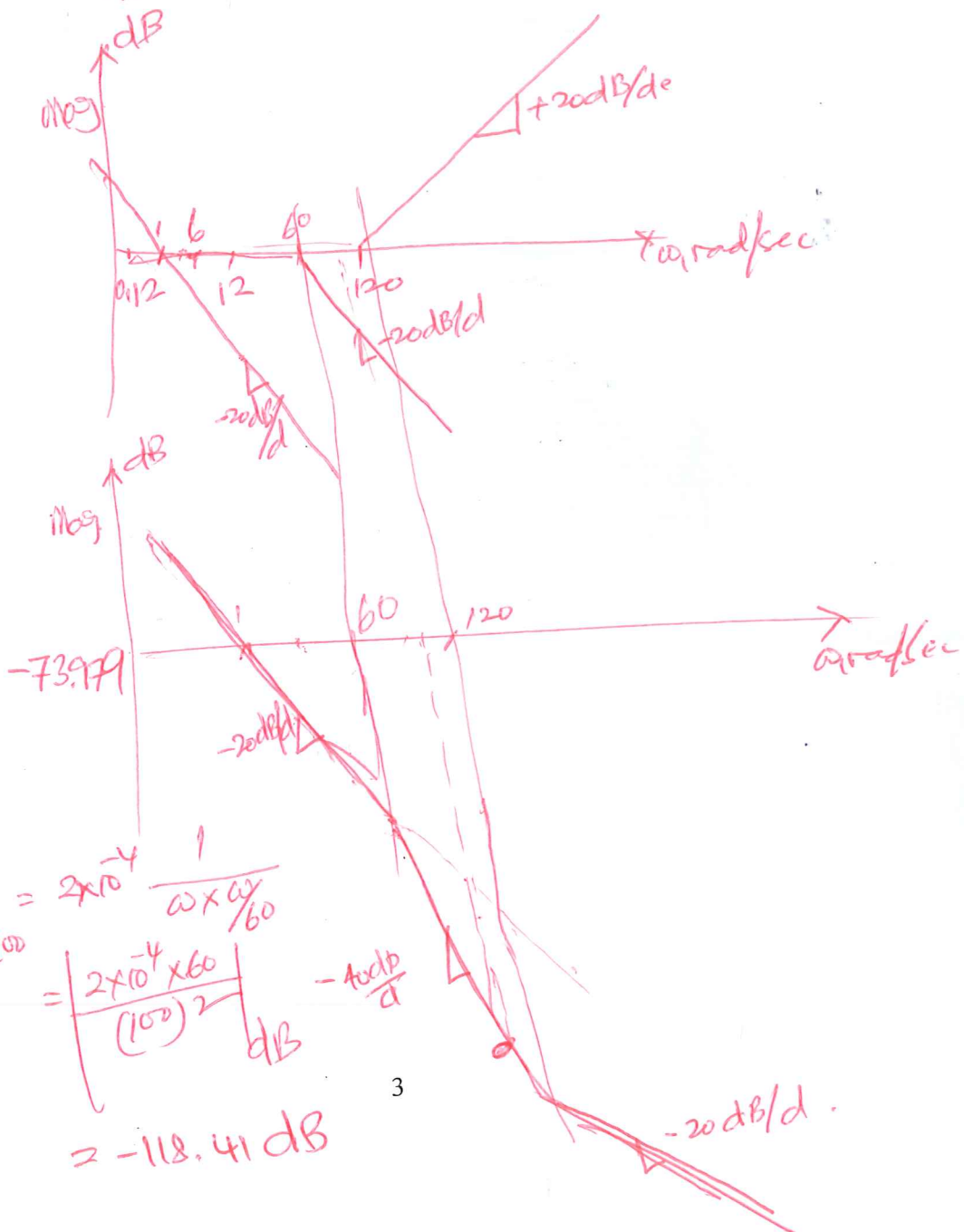
(b) Draw the Bode plot (magnitude and phase) of the following transfer function. (8 marks)

$$T(s) = \frac{10^{-4}(s + 120)}{s(s + 60)}$$

(c) Determine the magnitude and phase at  $s = 100$  rad/sec, using the solution from the graph of the question Q1.(b). (4 marks)

$$T(s) = \frac{10^{-4} \times 120 \left(1 + \frac{s}{120}\right)}{s \times 60 \left[1 + \frac{s}{60}\right]} = 2 \times 10^{-4} \times \frac{1 + \frac{s}{120}}{s \left(1 + \frac{s}{60}\right)}$$

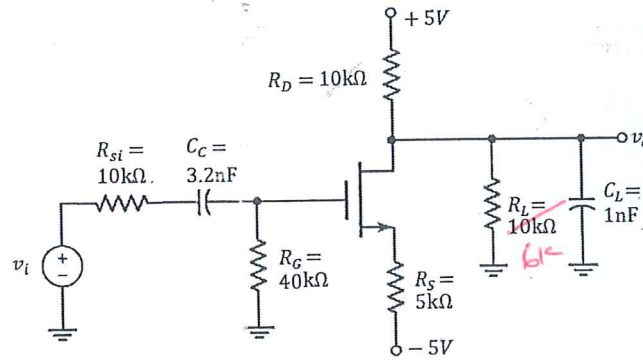
$$T_i/ds = 20 \log_{10} (2 \times 10^{-4}) = -73.979 \text{ dB}$$



$$\begin{aligned} |T(s)|_{\omega=100} &= 2 \times 10^{-4} \frac{1}{\omega \times \frac{\omega}{60}} \\ &= \frac{2 \times 10^{-4} \times 60}{(100)^2} \text{ dB} \\ &= -118.41 \text{ dB} \end{aligned}$$

**Q.2 [20 marks]**

A common source amplifier is shown in Fig. 2, with the following circuit component values  $R_{si} = 10k\Omega$ ,  $R_G = 40k\Omega$ ,  $R_S = 5k\Omega$ ,  $R_D = 10k\Omega$ ,  $R_L = 6k\Omega$ ,  $C_C = 3.2nF$  and  $C_L = 1nF$ . The MOSFET has a small - signal parameters,  $g_m = 40\text{ mA/V}$  and  $r_o = \infty$ .



**Fig. 2**

- (i) Draw the small signal equivalent circuit diagram of the circuit. (3 marks)
- (ii) Evaluate the midband voltage gain,  $A_v = \frac{v_o}{v_i}$  of the amplifier in dB scale. (7 marks)
- (iii) Determine the lower corner frequency,  $f_L$  (3 dB frequency) due to  $C_C$ . (4 marks)
- (iv) Determine the higher corner frequency  $f_H$ . (4 marks)
- (v) What is the bandwidth of the amplifier? (2 marks)

*Handwritten solution:*

Small signal equivalent circuit diagram showing the input signal  $v_i$  through  $10k$  resistor and  $3.2n$  capacitor to the gate. The gate is biased by  $R_G = 40k$ . The source is connected to ground through  $R_S = 5k$ . The drain is connected to  $+5V$  through  $R_D = 10k$ . The output  $v_o$  is taken from the drain, connected to a load resistor  $R_L = 6k$  and a load capacitor  $C_L = 1nF$ . The MOSFET parameters are  $g_m = 40\text{ mA/V}$  and  $r_o = \infty$ .

Calculations:

$$\tau_L = (R_{si} + R_G) C_C$$

$$= (10 + 40) k \times 3.2 nF$$

$$= 160 \mu s$$

$$f_L = \frac{1}{2\pi \tau_L} = 994.7 \text{ Hz}$$

$$R_L' = R_D \parallel R_L = 3.75 k$$

$$\tau_H = R_L' \parallel C_L$$

$$= 3.75 \mu s$$

$$f_H = \frac{1}{2\pi \tau_H} = 42.44 \text{ kHz}$$

$$BW = f_H - f_L = 41044.5 \text{ kHz}$$

$$A_{vA} = - \frac{g_m V_{gs} R_L}{(V_{gs} + g_m V_{gs} R_s)} = \frac{v_o}{v_{gs}}$$

$$= \frac{-g_m R_L}{(1 + g_m R_s)} = \frac{-40 \times 3.75k}{1 + 40 \times 15k}$$

$$= -0.746$$

$$A_v = A_{vA} \times \frac{R_L}{R_s + R_L}$$

$$= \textcircled{A_{vA}} -0.746 \times \frac{40}{10+50} = -0.597$$

$$A_v / \text{dB} = 20 \log 0.597 = \underline{\underline{-4.48 \text{ dB}}}$$

**Q.3 [20 marks]**

- (a) Draw the simplified high-frequency model of a bipolar junction transistor (BJT) and mention the name of each parameter. (2 marks)
- (b) Draw the simplified high-frequency small-signal short-circuit ( $R_C = 0$ ) equivalent diagram of the circuit as shown in Fig. 3(b), and determine the cutoff frequency  $f_T$  (step by step) for the transistor. (14 marks)
- (c) Assume that a BJT has the parameters,  $r_\pi = 1.5k\Omega$ ,  $g_m = \frac{20mA}{V}$ ,  $C_\pi = 20pF$  and  $C_\mu = 12pF$ , determine the  $f_\beta$  and  $f_T$  of the transistor. (4 marks)

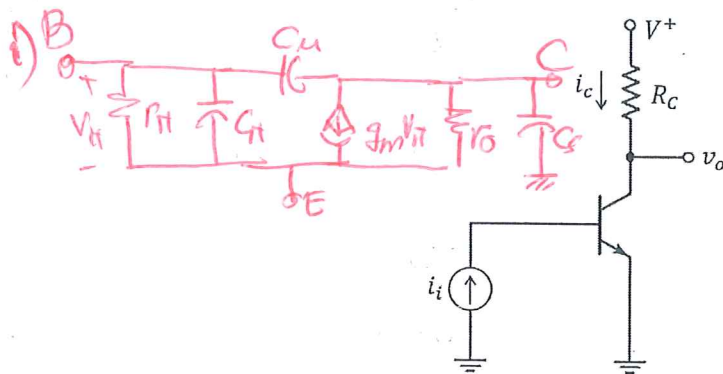


Fig. 3(b)

(i)

$$I_i = \frac{V_\pi}{r_\pi} + sC_\pi V_\pi + sC_\mu V_\pi = \frac{V_\pi}{r_\pi} [1 + s r_\pi (C_\pi + C_\mu)]$$

$$I_o = g_m V_\pi - V_\pi s C_\mu$$

$$\therefore A_{i_c} = \frac{I_o}{I_i} = \frac{(g_m - s C_\mu) V_\pi}{\frac{V_\pi}{r_\pi} [1 + s r_\pi (C_\pi + C_\mu)]} = \frac{g_m r_\pi - s C_\mu r_\pi}{1 + s r_\pi (C_\pi + C_\mu)}$$

$$\approx \frac{g_m r_\pi}{1 + s r_\pi (C_\pi + C_\mu)} = \frac{\beta}{1 + s r_\pi (C_\pi + C_\mu)}$$

[  $\because g_m r_\pi \gg \omega C_\mu r_\pi$  within freq of interest ]

$$|A_i| = 1 = \left| \frac{\beta}{1 + s r_\pi (C_\pi + C_\mu)} \right|$$

$$= \frac{\beta}{\sqrt{1 + [\omega_T r_\pi (C_\pi + C_\mu)]^2}} \approx \frac{\beta}{\omega_T r_\pi (C_\pi + C_\mu)}$$

$$\therefore \omega_T = \frac{\beta}{r_\pi (C_\pi + C_\mu)} \Rightarrow f_T = \frac{\beta}{2\pi r_\pi (C_\pi + C_\mu)} = \beta f_\beta =$$

$$f_T = \frac{\beta}{2\pi r_{\pi} (C_u + C_{\pi})}$$

$$= \frac{30}{2\pi \times 1.5k \times 32p} = 99.4718 \mu\text{Hz} \leftarrow$$

$$f_T = \beta f_B$$

$$\Rightarrow f_B = \frac{f_T}{\beta} = 3.3157 \text{ MHz} \leftarrow$$

$$\begin{aligned} \beta &= 20 \times 1.5 \\ &= 30 \\ &= \underline{\underline{30}} \end{aligned}$$