

Name: _____

Matric No: _____ Section: _____



الجامعة الإسلامية العالمية ماليزيا

INTERNATIONAL ISLAMIC UNIVERSITY MALAYSIA

MID-TERM EXAMINATION
SEMESTER II, 2017/2018 SESSION
KULLIYAH OF ENGINEERING

Programme : ENGINEERING Level of Study : UG 2
Time : 8:00 pm-10:00 pm Date : 25/10/2018
Duration : 2 Hours
Course Code : EECE 2313/ ECE 2133 Section(s) : 1, 2
Course Title : **Electronic Circuits**

This Question Paper consists of Six (6) Printed Pages (Including Cover and a blank page) with **Three (3)** Questions.

INSTRUCTION(S) TO CANDIDATES

DO NOT OPEN UNTIL YOU ARE ASKED TO DO SO

- Use only pen for writing the answer.
- Do not use your own paper sheet and no extra paper will be provided
- A total mark of this examination is 60.
- This examination is worth 30% of the total assessment.
- For drawing you may use a pencil
- Answer **ALL THREE(3)** questions
- Answer on the question paper

Any form of cheating or attempt to cheat is a serious offence which may lead to dismissal.

	Question 1	Question 2	Question 3	Total Marks
Marks	20	20	20	60
Marks Obtained				

Q.1 [20 marks]

- (a) Consider the circuit as shown in Fig. 1(a), derive the expression (step by step) for the voltage transfer function $T(s) = v_o(s)/v_i(s)$. Find the corner frequencies of the circuit if $R_s = 3 \text{ k}\Omega$, $R_p = 9 \text{ k}\Omega$ and $C_p = 10 \text{ nF}$ (6+2 marks)

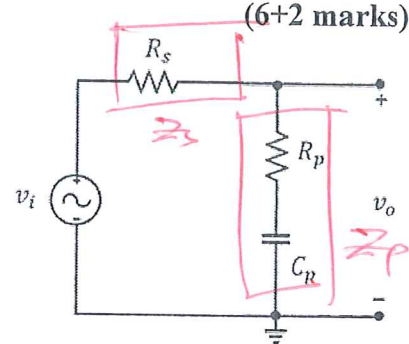


Fig. 1(a)

$$T(s) = \frac{V_o(s)}{V_i(s)} = \frac{Z_p}{Z_p + Z_s}$$

$$Z_p = R_p + \frac{1}{sC_p} = \frac{1 + sC_p R_p}{sC_p}$$

$$Z_p + Z_s = R_p + \frac{1}{sC_p} + R_s = \frac{1 + sC_p(R_p + R_s)}{sC_p}$$

$$\therefore T(s) = \frac{\frac{(1 + sC_p R_p)/sC_p}{sC_p}}{\frac{1 + sC_p(R_p + R_s)}{sC_p}} = \frac{1 + sC_p R_p}{1 + sC_p(R_p + R_s)}$$

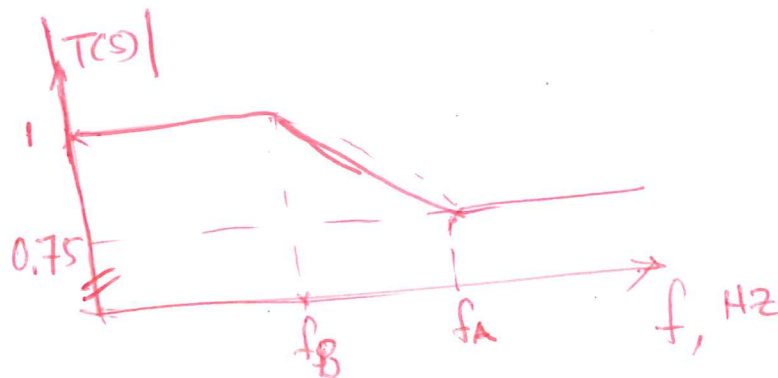
$$= \frac{1 + s\tau_A}{1 + s\tau_B} \quad \leftarrow$$

$$\tau_A = C_p R_p = 9 \text{ k} \times 10 \text{ n} = 90 \text{ n s}$$

$$\therefore f_A = \frac{1}{2\pi\tau_A} = 1.77 \text{ kHz}$$

$$\tau_B = C_p(R_p + R_s) = 10 \text{ n} \times (9 \text{ k} + 3 \text{ k}) = 120 \text{ n s}$$

$$f_B = \frac{1}{2\pi\tau_B} = 1.326 \text{ kHz}$$



(b) Draw the Bode plot (magnitude and phase) of the following transfer function.

(8 marks)

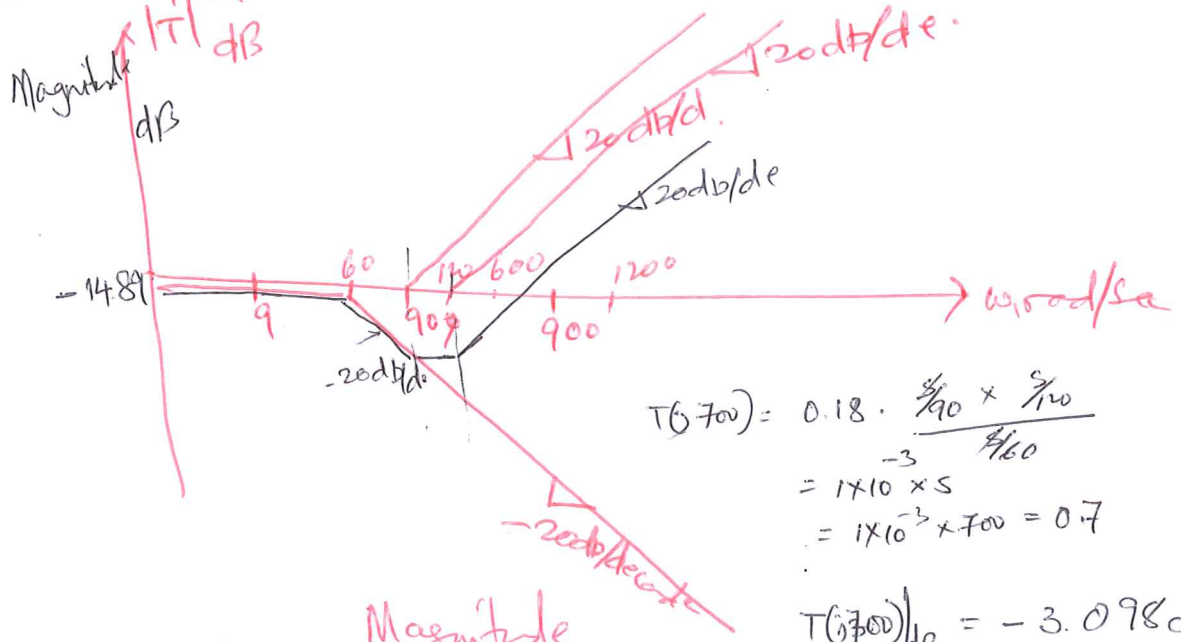
$$T(s) = \frac{10^{-3}(s + 90)(s + 120)}{(s + 60)}$$

(c) Determine the magnitude and phase at $s = 700$ rad/Sec, using the solution from the graph of the question Q1.(b).

(4 marks)

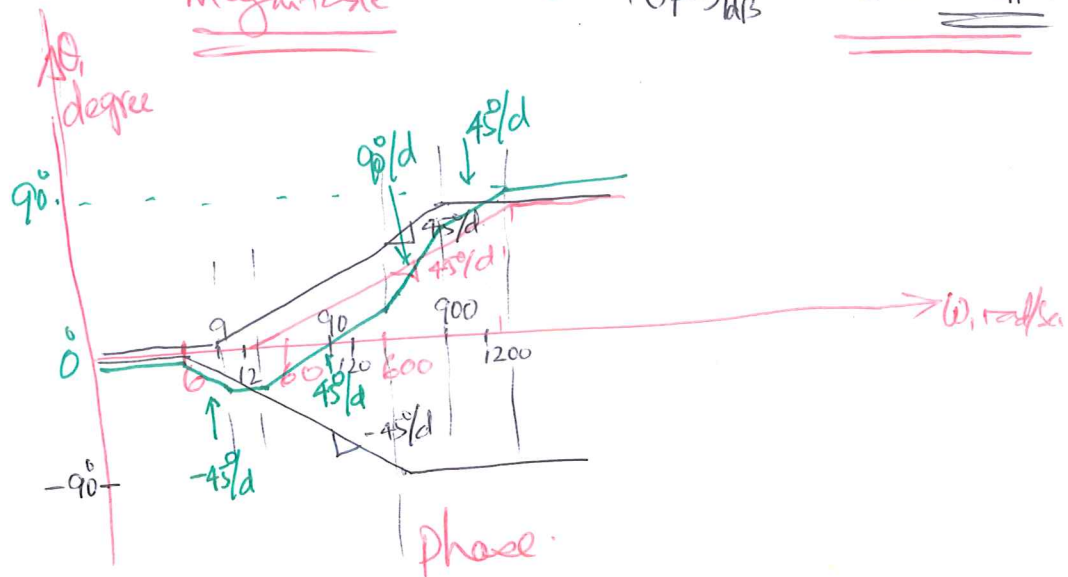
$$T(s) = \frac{10^{-3} \cdot 90 \cdot 120 \cdot (1 + \frac{s}{120})}{60 \cdot [1 + \frac{s}{60}]} = 0.18 \frac{(1 + \frac{s}{120})(1 + \frac{s}{90})}{(1 + \frac{s}{60})}$$

$$0.18 \rightarrow -14.89 \text{ dB}$$



$$T(700) = 0.18 \cdot \frac{90 \times 700}{60} = 1 \times 10^{-3} \times 700 = 0.7$$

$$T(700)_{dB} = -3.098 \text{ dB}$$



$$\rightarrow \theta = \tan^{-1} \frac{\omega}{90} + \tan^{-1} \frac{\omega}{120} - \tan^{-1} \frac{\omega}{60}$$

$$= 82.67 + 80.27 - 85.1 = 77.83^\circ$$

$\omega = 700 \text{ rad/sec}$

Q.2 [20 marks]

A common source amplifier is shown in **Fig. 2**, with the following circuit component values $R_{si} = 10\text{k}\Omega$, $R_G = 40\text{k}\Omega$, $R_S = 5\text{k}\Omega$, $R_D = 10\text{k}\Omega$, $R_L = 6\text{k}\Omega$, Design the circuit for lower corner frequency $f_L = 250\text{Hz}$ and higher corner frequency $f_H = 100\text{kHz}$. The MOSFET has a small-signal parameters, $g_m = 45 \text{ mA/V}$ and $r_o = \infty$.

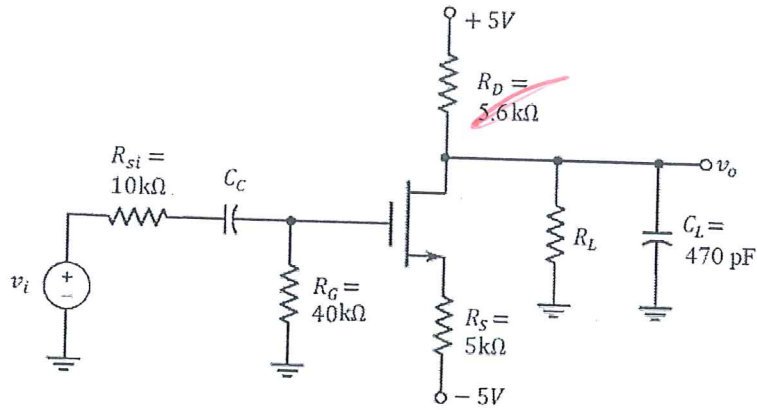


Fig. 2

- (i) Draw the small signal equivalent circuit diagram of the circuit (3 marks)
- (ii) Evaluate the midband gain, $A_v = v_o/v_i$ of the amplifier in *dB* scale (7 marks)
- (iii) Determine the coupling capacitance C_C (3 marks)
- (iv) Determine the load resistance R_L (7 marks)

Q.3 [20 marks]

A common emitter amplifier is shown in **Fig. 3** that operates at very high frequencies. The transistor parameters are: $g_m = 40\text{mA/V}$, $r_\pi = 2.5\text{k}\Omega$ and $r_o = 25\text{k}\Omega$, $C_\pi = 8\text{pF}$ and $C_\mu = 2\text{pF}$. Assume that $C_{C2} = C_E = \infty$ and $C_{C1} = 3.3\mu\text{F}$.

- i) Draw the simplified high-frequency small signal equivalent circuit diagram and Miller equivalent circuit diagram (3 marks)
- ii) Determine the midband gain $A_v = v_o/v_i$ (7 marks)
- iii) Write the expression of Miller capacitance and find its value (3 marks)
- iv) Evaluate the upper 3dB frequency (f_H) considering Miller capacitance and without considering Miller capacitance (2 marks)
- v) Evaluate the lower 3dB frequency (f_L) (5 marks)

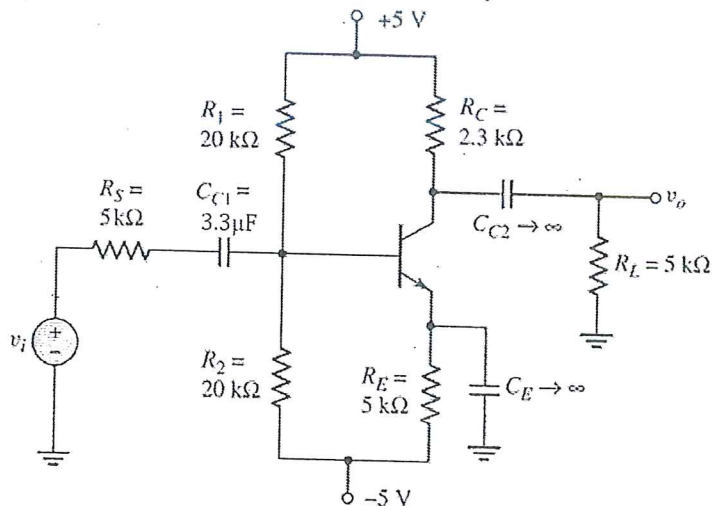
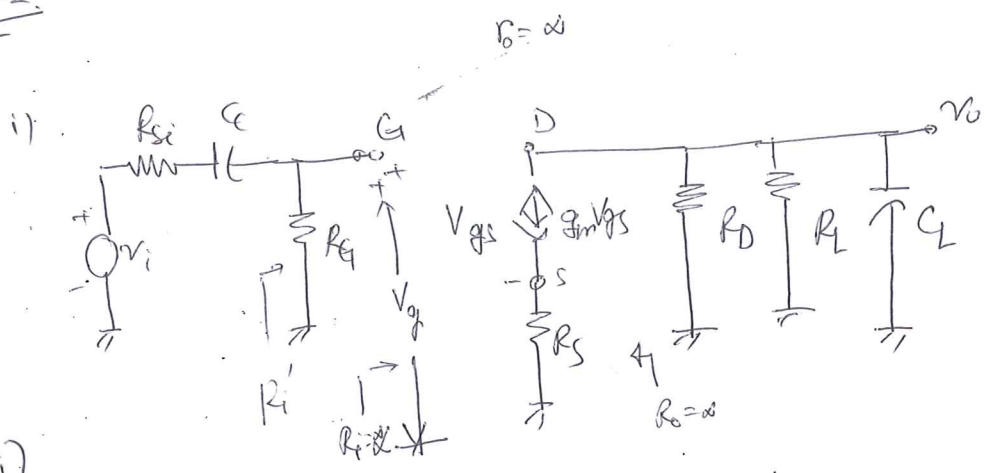


Fig. 3

Q.2



(ii)

$$A_{V_{midband}} = \frac{v_o}{v_{gs}} = \frac{-g_m v_{gs} R_L'}{v_{gs} + g_m v_{gs} R_s} \quad R_L' = R_D || R_L$$

$$= \frac{-g_m R_L'}{(1 + g_m R_s)} \quad R_L = ?$$

(iii)

$$A_{V_{in}} = A_{V_{mid}} \times \frac{R_i}{R_i + R_{si}} \quad R_i = R_G$$

$$= A_{V_{mid}} \times \frac{R_G}{R_G + R_{si}} = A_{V_{mid}} \times \frac{40}{40 + 10} = 0.8 A_{V_{mid}}$$

(iii)

$$f_L = \frac{1}{2\pi(R_{si} + R_i)C_c} = 12.73 \text{ nF} \leftarrow$$

$$C_c = \frac{1}{2\pi f_L (R_{si} + R_G)} = \frac{1}{2\pi \times 250 \times (40 + 10) \times 10^3}$$

(iv)

$$C_L = \frac{1}{2\pi f_H R_L'} \quad R_L' = \frac{1}{2\pi C_L f_H} = \frac{1}{2\pi \times 470 \times 100 \times 10^3}$$

$$= 3.386 \text{ k}\Omega = R_D || R_L$$

$$R_L = 5.12 \text{ k}\Omega \star$$

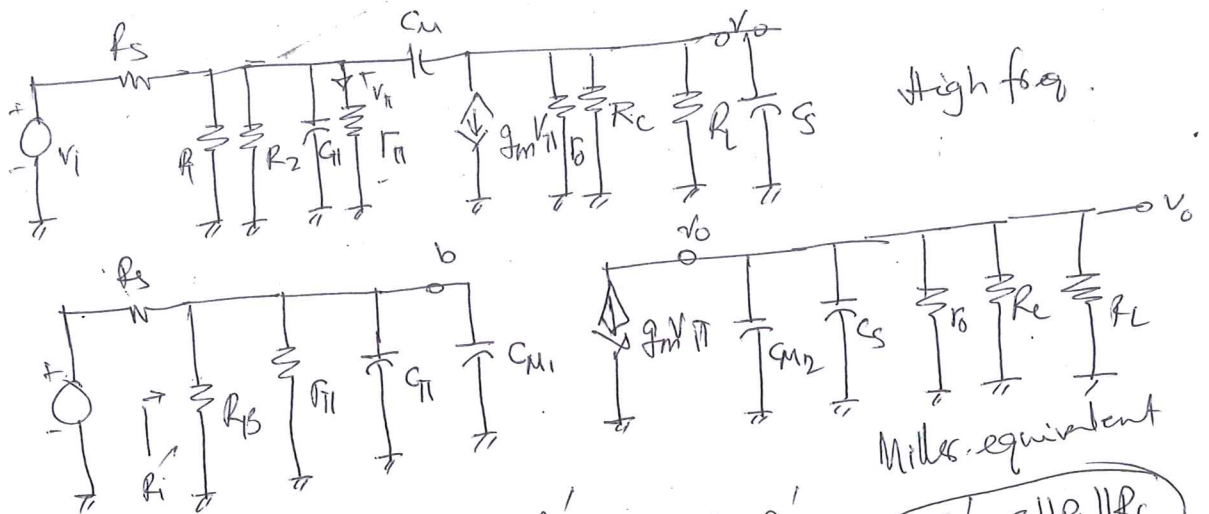
$$\frac{1}{R_L} + \frac{1}{R_D} = \frac{1}{R_L'} = 295.3 \times 10^{-6}$$

(iii)

$$A_{V_{mid}} = \frac{-g_m R_L'}{1 + g_m R_s} = \frac{45 \times 3.386 \times 10^3}{1 + 45 \times 5 \times 10^3} = 0.674$$

$$A_V = 0.8 \times 0.674 \approx 0.54 \text{ V/V} = \underline{\underline{-5.36 \text{ dB}}}$$

Q.3



High freq.

Miller equivalent

$$R_L' = 6 \parallel 2.3 \parallel 5 = 1.482 \text{ k}\Omega$$

ii)

$$A_{vA/m} = + \frac{v_o}{v_b} = \frac{-g_m V_{th} R_L'}{V_{th}} = -g_m R_L'$$

$$= -40 \text{ m} \times 1.482 \text{ k} = -59.28 \text{ V/V}$$

$$R_i' = R_1 \parallel R_2 \parallel R_{th} = 20 \parallel 20 \parallel 2.5 = 2 \text{ k}\Omega$$

$$A_{vm} = A_{vA} \times \frac{R_i'}{R_i' + R_s} = -59.28 \times \frac{2}{2+5}$$

$$= -16.94 \text{ V/V} = 24.57 \text{ dB}$$

(iii)

$$C_{M1} = C_u (1 + |A_{vA}|) = 2 \text{ pF} (1 + 59.28) = 120.56 \text{ pF}$$

$$C_{M2} = C_u (1 + \frac{1}{|A_{vA}|}) = 2 \text{ pF} (1 + \frac{1}{59.28}) = 2.034 \text{ pF}$$

can be neglected compared to \$C_{M1}\$ effect.

(iv)

$$f_H = \frac{1}{2\pi C_{eq} R_{eq}}$$

$$R_{eq} = R_s \parallel R_1 \parallel R_2 \parallel R_{th} = 5 \parallel 20 \parallel 20 \parallel 2.5 = 1.4286 \text{ k}\Omega$$

$$C_{eq} = \frac{1}{2\pi \times 1.4286 \text{ k} (8 + 120.56 \text{ pF})}$$

$$f_{H \text{ Miller}} = \frac{1}{2\pi R_{eq} (C_{\pi} + C_{M1})} = \frac{1}{2\pi \times 1.4286 \text{ k} \times 8 \text{ pF}} = 13.93 \text{ MHz}$$

$$f_{H \text{ (without Miller)}} = \frac{1}{2\pi R_{eq} C_{\pi}} = \frac{1}{2\pi \times 1.4286 \text{ k} \times 3.3 \text{ nF}} = 6.89 \text{ Hz}$$

(v)

$$f_H = \frac{1}{2\pi (R_s + R_i') C_u} = \frac{1}{2\pi \times (5+2) \text{ k} \times 3.3 \text{ nF}}$$