

MODEL ANSWER

Name: \_\_\_\_\_

Matric No: \_\_\_\_\_ Section: \_\_\_\_\_



الجامعة الإسلامية العالمية ماليزيا

INTERNATIONAL ISLAMIC UNIVERSITY MALAYSIA

MID-TERM EXAMINATION  
SEMESTER II, 2012/2013 SESSION  
KULLIYAH OF ENGINEERING

Programme : ENGINEERING Level of Study : UG 2  
Time : 8:00pm-10:00 pm Date : 22/03/2013  
Duration : 2 Hours  
Course Code : ECE 2133 Section(s) : 1 & 2  
Course Title : **Electronic Circuits**

This Question Paper consists of **Seven (7)** Printed Pages (Including Cover and a blank page) with **Two (2)** Questions.

**INSTRUCTION(S) TO CANDIDATES**

DO NOT OPEN UNTIL YOU ARE ASKED TO DO SO

- Use only pen for writing answer.
- Do not use your own sheet.
- Useful formulas and values are given in page 7.
- A total mark of this examination is 60.
- This examination is worth 30% of the total assessment.
- For drawing you may use pencil
- Answer **ALL** questions.

**Any form of cheating or attempt to cheat is a serious offence which may lead to dismissal.**

	Question 1a	Question 1b	Question 2a	Question 2b	Total Marks
Marks	15	15	15	15	60
Marks Obtained					

**Q.1 [30 marks]**

(a) Determine the Q-point values ( $I_{DQ}$  and  $V_{DSQ}$ ) for the circuit shown in Fig. 1(a). The transistor parameters are  $V_{TN} = 1.2$  V,  $K_n = 0.5$  mA/V<sup>2</sup> and  $\lambda = 0$ . Also calculate the small signal circuit parameters. (15 marks)

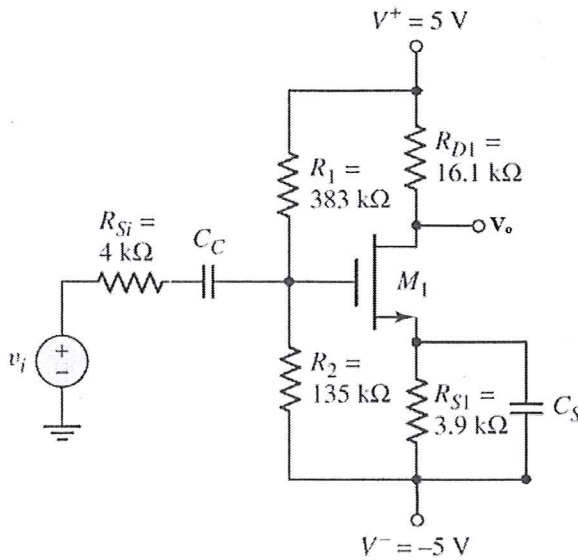
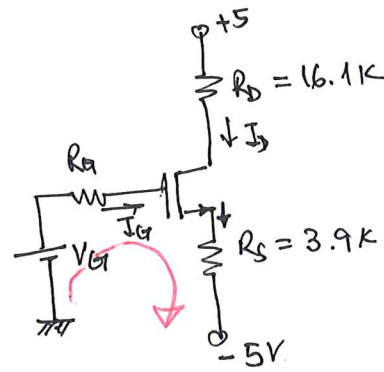


Fig. 1(a)

$$R_G = R_1 \parallel R_2 = 383 \parallel 135 = 99.8167 \text{ k}\Omega$$



$$5 = I_D (R_D + R_S) + V_{DS} - 5$$

$$\therefore V_{DS} = 5 + 5 - I_D (R_D + R_S)$$

$$= 10 - 0.19845 \times 20$$

$$= 6.031 \text{ V} \quad \leftarrow$$

$$V_G = \frac{(5+5) \times 135}{135+383} \times 135 - 5 = \frac{10 \times 135}{518} - 5 = -2.394 \text{ V} \leftarrow$$

$$V_G = I_D R_G + V_{GS} + R_S I_D - 5 = V_{GS} + R_S K_n (V_{GS} - V_T)^2 - 5$$

$$-2.394 = V_{GS} + 3.9 \text{ k} \times 0.5 \text{ mA/V}^2 (V_{GS} - 1.2)^2 - 5$$

$$= V_{GS} + 1.95 (V_{GS}^2 - 2.4 V_{GS} + 1.44) - 5$$

$$= 1.95 V_{GS}^2 - 3.68 V_{GS} + 2.808 - 5$$

$$\therefore 1.95 V_{GS}^2 - 3.68 V_{GS} + 0.202 = 0 \quad \leftarrow$$

$$V_{GS} = 1.83 \text{ V}, 0.056 \text{ V}$$

choosing  $V_{GS} = 1.83 > V_T$ .

$$I_D = K_n (V_{GS} - V_T)^2 = 0.5 (1.83 - 1.2)^2$$

$$= 0.19845 \text{ mA} \quad \leftarrow$$

$$g_m = 2 \sqrt{K_n I_{DQ}} = 2 \sqrt{0.5 \times 0.19845}$$

$$= 0.63 \text{ mA/V} \quad \leftarrow$$

$$r_o = \frac{1}{\lambda I_{DQ}} = \infty \quad \leftarrow$$

- (b) For the circuit shown in Fig 1(b), design a bias stable circuit such that  $I_{CQ} = 0.8 \text{ mA}$  and  $V_{CEQ} = 5.0 \text{ V}$  assuming  $\beta = 100$ . Also find small signal parameters  $g_m$ ,  $r_\pi$  and  $r_o$  if  $V_A = 100$ . [Hints: determine the values of  $R_C$ ,  $R_1$  and  $R_2$ ] (15 marks)

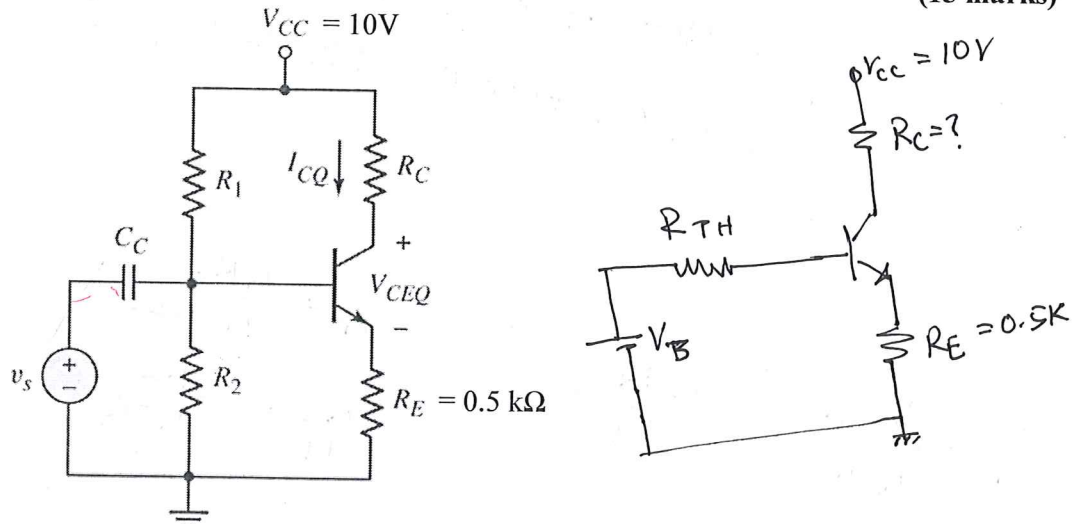


Fig. 1 (b)

For bias stable,  $R_{TH} = R_B = 0.1(1 + \beta) R_E = 0.1(1 + 100) \times 0.5 \text{ k}\Omega = 5.05 \text{ k}\Omega$

$$V_{CC} = I_C R_C + V_{CEQ} + I_E R_E$$

$$= I_C R_C + V_{CEQ} + (1 + \beta) I_B R_E$$

$$\therefore I_C R_C = V_{CC} - V_{CEQ} - \frac{(1 + \beta) I_C R_E}{\beta}$$

$$= 10 - 5 - \frac{101}{100} \times 0.5 \times 0.8 \text{ m}$$

$$= 4.5960$$

$$\therefore R_C = 5.745 \text{ k}\Omega \quad \leftarrow$$

$$\begin{aligned}
 V_B &= I_B R_B + V_{BE} + I_E R_E \\
 &= \frac{I_C R_B}{\beta} + V_{BE} + \frac{(1+\beta) I_C R_E}{\beta} \\
 &= \frac{0.8}{100} \times 5.05 + 0.7 + \frac{101 \times 0.8 \times 0.5}{100} \\
 &= 0.0404 + 0.7 + 0.404 = 1.1444 \text{ V} \leftarrow
 \end{aligned}$$

$$V_B = \frac{V_C R_2}{R_1 + R_2} = \frac{V_C R_B}{R_1}$$

$$R_B = \frac{R_1 R_2}{R_1 + R_2}$$

$$\frac{R_B}{R_1} = \frac{R_2}{R_1 + R_2}$$

$$\begin{aligned}
 \therefore R_1 &= \frac{V_C \cdot R_B}{V_B} \\
 &= \frac{10}{1.1444} \times 5.05 \text{ k} = 44.128 \text{ k}\Omega \leftarrow
 \end{aligned}$$

$$R_2 = \frac{1}{\frac{1}{R_B} - \frac{1}{R_1}} = 5.70 \text{ k}\Omega \leftarrow$$

$$\frac{1}{R_B} = \frac{1}{R_1} + \frac{1}{R_2}$$

$$g_m = \frac{I_{CQ}}{0.026} = \frac{0.8}{0.026} = 30.77 \text{ mA/V}$$

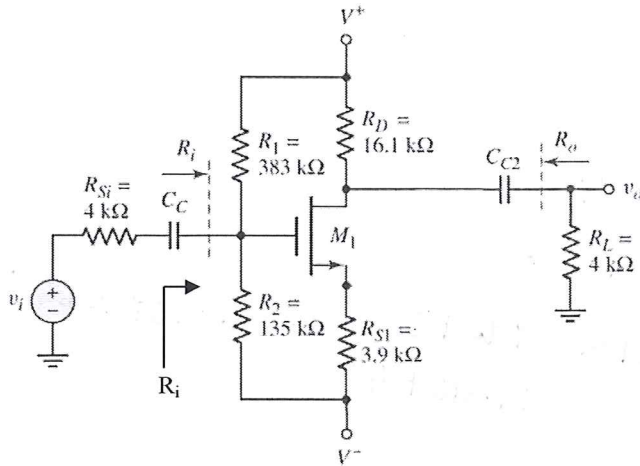
$$\beta = g_m r_{\pi}$$

$$r_{\pi} = \frac{\beta}{g_m} = \frac{100}{30.77} = 3.25 \text{ k}\Omega \leftarrow$$

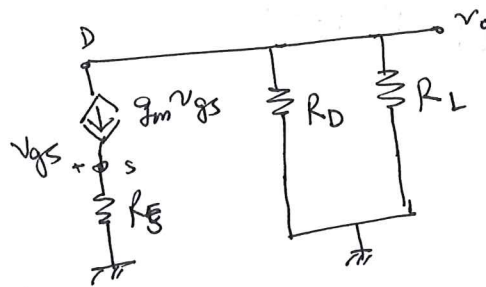
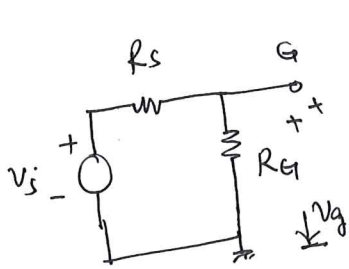
$$r_o = \frac{V_A}{I_{CQ}} = \frac{80}{0.8} = 100 \text{ k}\Omega \leftarrow$$

**Q.2 [20 marks]**

(a) Draw the small signal equivalent circuit diagram of the MOSFET amplifier circuit shown in Fig. 2(a) and find the input resistance  $R_i$  and the voltage gain  $A_v = \frac{v_o}{v_s}$  of the amplifier. The transistor parameters are  $g_m = 0.65 \text{ mA/V}$  and  $\lambda = 0$ . (15 marks)



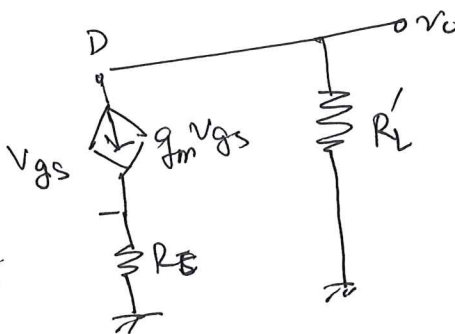
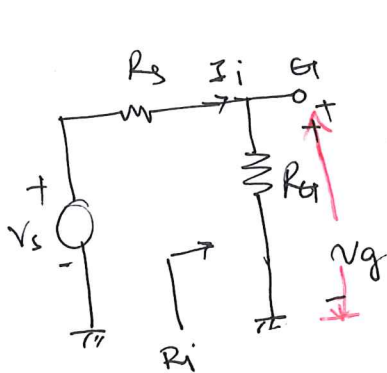
$\lambda = 0 \quad r_o = \infty$



$$R_i = \frac{v_g}{I_i} = \frac{R_G I_i}{I_i}$$

$$= R_G = R_1 \parallel R_2$$

$$= 99.816 \text{ k}\Omega$$



$$v_o = -g_m v_{gs} \times R_L'$$

$$\therefore A_{vA} = \frac{v_o}{v_s} = \frac{-g_m v_{gs} R_L'}{v_{gs} (1 + g_m R_E)}$$

$$= -\frac{g_m R_L'}{1 + g_m R_E}$$

$$v_g = v_{gs} + g_m v_{gs} R_E$$

$$= v_{gs} (1 + g_m R_E)$$

$$A_{v_A} = \frac{-g_m R'_L}{1 + g_m R_s}$$

$$R'_L = 16.1 \parallel 4 \\ = 3.204 \text{ k}\Omega$$

$$= \frac{-0.69 \times 3.204}{1 + 0.69 \times 3.9} = -0.5891$$

$$\therefore A_v = \frac{v_o}{v_s} = A_{v_A} \times \frac{R_i}{R_i + R_s} \\ = -0.5891 \times \frac{99.816}{99.816 + 4} = -0.5664 \quad \leftarrow$$

$$R_i = 99.816 \text{ k}\Omega \quad \leftarrow$$

(b) Draw the small signal equivalent circuit diagram of the BJT amplifier circuit

shown in Fig. 2(b) and find the current gain  $A_i = \frac{I_o}{I_i}$  and output resistance  $R_o$

where  $I_o$  is the current flowing through  $R_L$ . Given that  $R_1 = 300 \text{ k}\Omega$ ,  $R_2 = 500 \text{ k}\Omega$ ,  $R_E = 20.0 \text{ k}\Omega$  and  $I_D = 0.25 \text{ mA}$ . The transistor parameters are  $\beta = 100$ ,  $V_{BE(on)} = 0.7 \text{ V}$  and  $V_A = 80$ . (15 marks)

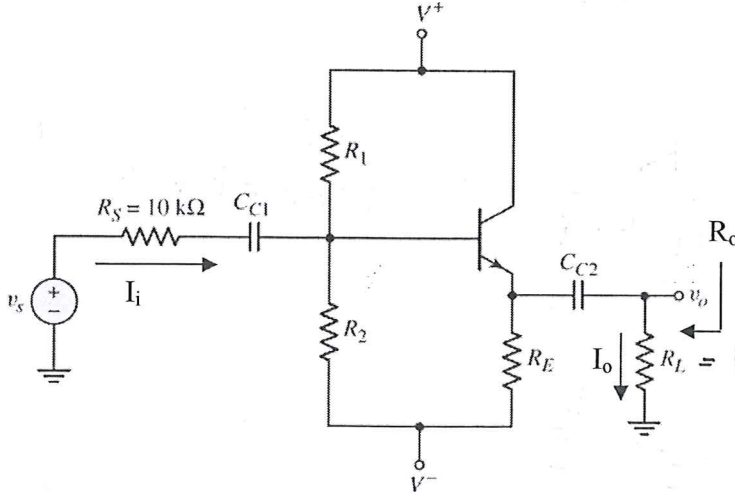
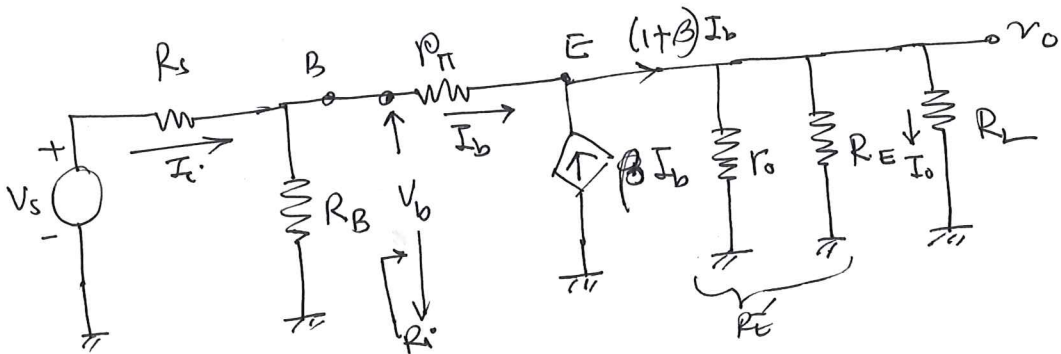


Fig. 2(b)

$$r_o = \frac{V_A}{I_C} = \frac{80}{0.25 \text{ mA}} = 320 \text{ k}\Omega \leftarrow$$

$$r_{\pi} = \frac{\beta V_T}{I_C} = \frac{100 \times 0.026}{0.25} = 10.4 \text{ k}\Omega \leftarrow$$

$$R'_L = r_o \parallel R_E \parallel R_L = 320 \parallel 20 \parallel 5 = 3.95 \text{ k}\Omega$$

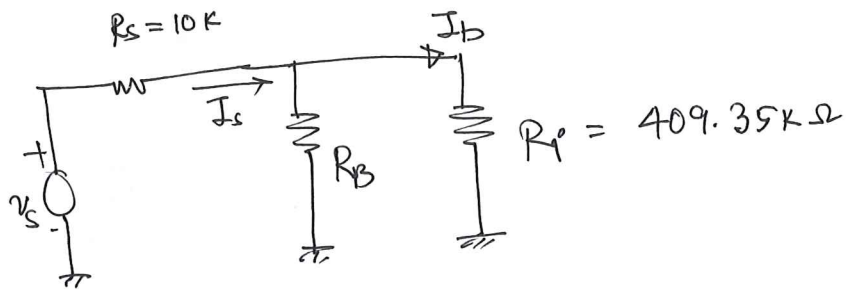


$$R'_E = r_o \parallel R_E = 320 \parallel 20 = 18.82 \text{ k}\Omega$$

$$I_o = \frac{I_b (1 + \beta) \times R'_E}{R_L + R'_E} = 100 I_b \times \frac{18.82}{5 + 18.82} = 79.787 I_b$$

$$R_i = \frac{V_b}{I_b} = \frac{r_{\pi} I_b + I_b (1 + \beta) R'_E}{I_b} = r_{\pi} + (1 + \beta) R'_E = 10.4 + 101 \times 3.95 = 409.35 \text{ k}\Omega \leftarrow$$

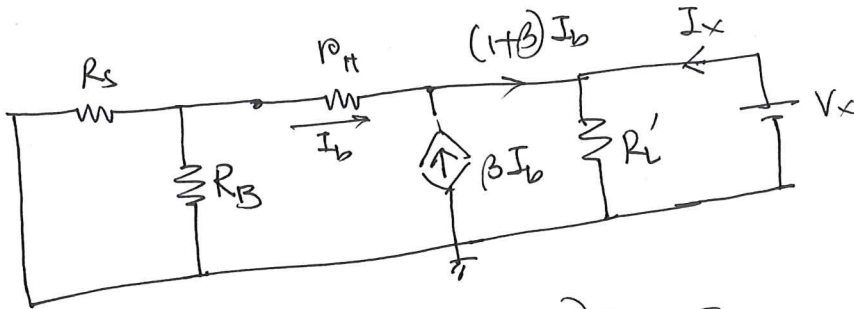
$$R_B = R_1 \parallel R_2 = 300 \parallel 500 = 187.5 \text{ k}$$



$$\frac{I_b}{I_s} = \frac{R_B}{R_B + R_i} = \frac{187.5}{187.5 + 409.35} = 0.3141$$

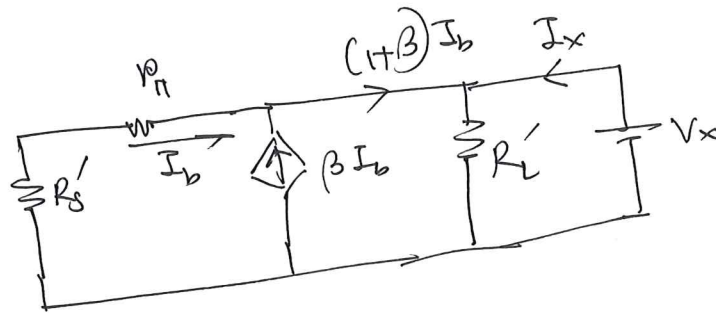
$$\therefore A_I^2 \frac{I_o}{I_s} = \frac{I_o}{I_b} \cdot \frac{I_b}{I_s} = 79.787 \times 0.3141 = 25.065 \text{ A}$$



$R_0 = ?$ 

$$R'_S = R_B \parallel R_s = 187.5 \parallel 10 = 9.494 \text{ k}\Omega$$

$$r_{\pi} = 10.4 \text{ k}\Omega$$



$$I_x = \frac{V_x}{R'_L} - (1+\beta)I_b$$

$$I_b = - \frac{V_x}{r_{\pi} + R'_S}$$

$$= \frac{V_x}{R'_L} + \frac{(1+\beta)V_x}{r_{\pi} + R'_S} = \frac{V_x}{R'_L} + \frac{V_x}{\frac{r_{\pi} + R'_S}{(1+\beta)}}$$

$$\therefore \frac{1}{R_0} = \frac{1}{R'_L} + \frac{1}{\frac{r_{\pi} + R'_S}{(1+\beta)}} \quad \therefore R_0 = R'_L \parallel \left( \frac{r_{\pi} + R'_S}{1+\beta} \right)$$

$$\therefore \frac{r_{\pi} + R'_S}{1+\beta} = \frac{10.4 + 9.494}{1+100} = 0.19696 \text{ k}\Omega$$

$$\therefore R_0 = 3.95 \text{ k}\Omega \parallel 0.19696 \text{ k}\Omega = 0.1876 \text{ k}\Omega$$

$$= 187.6 \Omega \leftarrow$$

**Useful formulas****For BJT**For bias stable  $R_B = 0.1(1 + \beta)R_E$ 

$$g_m = \frac{I_{CQ}}{V_T}$$

$$r_\pi = \frac{\beta V_T}{I_{CQ}}$$

$$r_o = \frac{V_A}{I_{CQ}}$$

$$V_T = 0.026 \text{ V}$$

$$V_{BE(\text{on})} = 0.7 \text{ V}$$

**For MOSFET**

$$g_m = 2\sqrt{k_n I_{DQ}}$$

$$r_o = \frac{1}{\lambda I_{DQ}}$$

*Large values of capacitors are connected to the circuits as coupling and bypass capacitors.*